

FDDS10H04A_F085A

Smart High Side Switch

Features

- Short circuit protection with latch
- Current limitation
- Overload protection
- Thermal shutdown with restart.
- Overvoltage protection(including load dump)
- Loss of ground protection
- Loss of supply protection(with external diode for charged inductive load)
- Very low standby current
- Fast demagnetization of inductive loads
- ESD protection
- Optimized static electromagnetic compatibility
- Diagnostic function - Proportional load current sense(with defined fault signal in case of overload operation, overtemperature shutdown and/or short circuit shutdown)
- Qualified to AEC

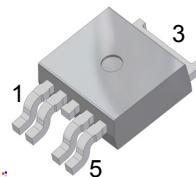
Typical Applications

- Power switch with current sense diagnostic feedback for DC ground loads
- All types of resistive, inductive, and capacitive loads
- Replace electromechanical relays, fuses and discrete circuits

Description

N channel power FET with charge pump, current controlled input and diagnostic feedback with load current sense, integrated in smart Trench chip on chip technology. Provides embedded protective functions.

TO252-5L



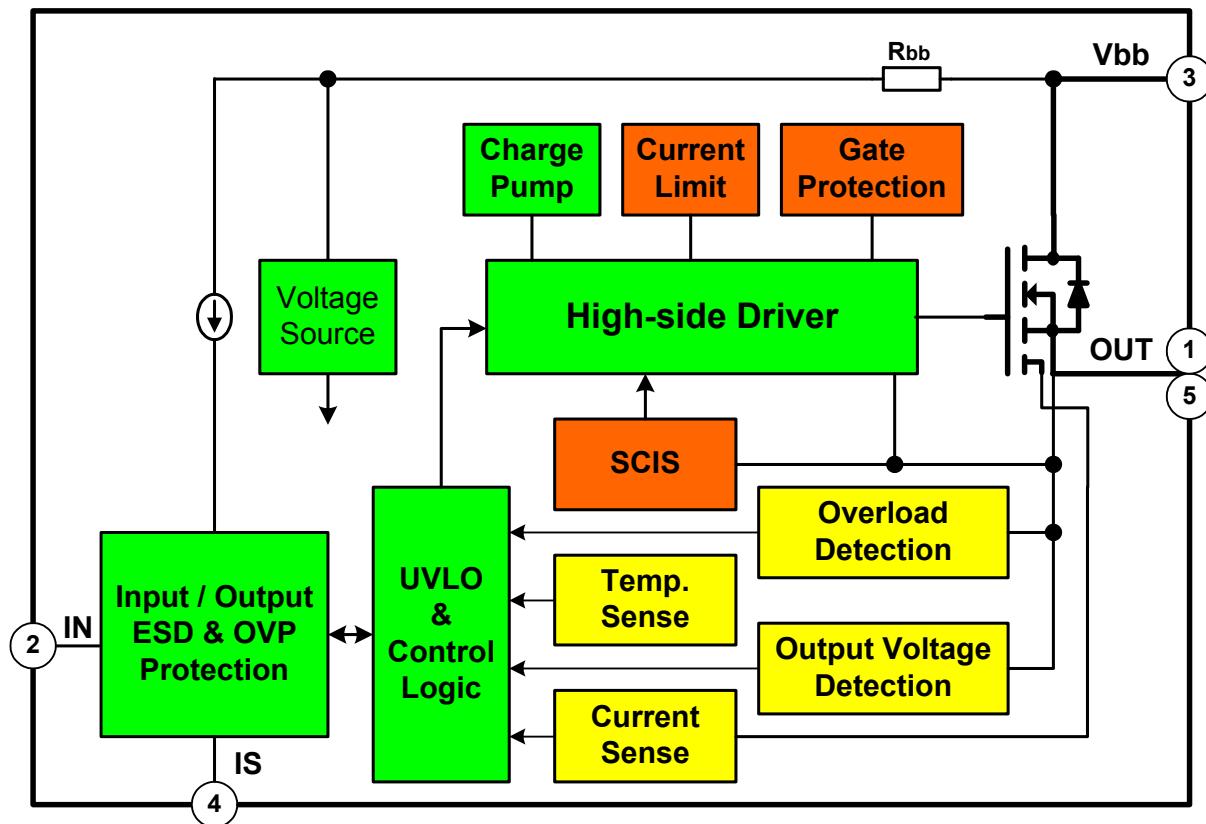
Ordering Information

Part Number	Package	Operating Temperature	Eco Status	Packing Method
FDDS10H04A_F085A	TO252-5L	-40 °C - 150 °C	RoHS	Tape & Reel



For Fairchild's definition of "green" Eco Status, please visit: http://www.fairchildsemi.com/company/green/rohs_green.html.

Block Diagrams



Pin Definitions

Pin Number	Pin Name	I/O	Pin Function Description
1	OUT	A	Output to loads; Pin 1 and 5 must be externally shorted
2	IN	A	Input: activates the power switch if shorted to ground
3	Vbb	P	Supply voltage; Pin 3 and TAB are internally shorted
4	IS	A	Sense output ; Diagnostic feedback; Provides at normal operation a sense current proportion to the load current; in case of overload, over temperature and/or short circuit a defined current is provided
5	OUT	A	Output to loads; Pin 1 and 5 must be externally shorted

Absolute Maximum Ratings

At $T_j=25^\circ\text{C}$ unless otherwise specified.

Parameter	Symbol	Values	Unit
Supply voltage ⁴⁾	V_{bb}	38	V
Supply voltage for full short circuit protection ¹⁾	V_{bb}	30	V
Load dump protection $V_{LoadDump} = U_A + V_S$, $U_A=13.5\text{V}$ $R_I=2\Omega$, $R_L=1\Omega$, $t_d=400\text{ms}$, $I_N=\text{Low or High}$	$V_{LoadDump}$ ²⁾	45	V
Load current (Short-circuit current)	I_L	Self-limited	V
Operating temperature range	T_j	-40 ~ 150	$^\circ\text{C}$
Storage temperature range	T_{stg}	-55 ~ 150	$^\circ\text{C}$
Power Dissipation(DC)	P_{tot}	59	W
Inductive load switch-off energy dissipation ³⁾ Single pulse, $I_L=10.7\text{A}$, $L=5\text{mH}$, $V_{bb}=12\text{V}$, $T_j=150^\circ\text{C}$	EAS	288	mJ
Electrostatic discharge capability (ESD) (Human Body Model)	I_S	V_{ESD}	2.8 KV
	I_N	V_{ESD}	3 KV
	VBB, Output	V_{ESD}	5 KV
Current through input pin(DC)	I_{IN}	+15, -120	mA
Current through current sense pin(DC)	I_{IS}	+15, -120	mA
Input voltage slew rate $V_{bb} \leq 16\text{V}$	dV_{bb}/dt		self-limited
Input voltage slew rate $V_{bb} > 16\text{V}$ ⁴⁾	dV_{bb}/dt		20 V/us

Note:

1) Short circuit is defined as a combination of remaining resistances and inductances. See schematic on page11.

2) VLoad dump is setup without the DUT connected to the generator

3) See also diagram on page 11.

4) See also on page 7. Slew rate limitation can be achieved by means of using a series resistor R_{IN} in the input path. This resistor is also required for reverse operation. See also page 10.

Electrical Characteristics

At $T_j=25^\circ\text{C}$, $V_{bb}=12\text{V}$ unless otherwise specified.

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
Thermal Characteristics						
Thermal resistance	$R_{thJC}^{(5)}$ R_{thJA}	(junction to case) (junction to ambient) device on PCB ⁽⁶⁾ , SMD version only	- - -	- 83 45	1.0 - 55	K/W
Load Switching Capability and Characteristics						
On-state resistance(Pin3 to pin 1,5)	R_{ON}	$V_{IN}=0, V_{bb}=5.5\text{V}, I_L=7.5\text{A}, T_j=25^\circ\text{C}$ $V_{IN}=0, V_{bb}=5.5\text{V}, I_L=7.5\text{A}, T_j=150^\circ\text{C}$ $V_{IN}=0, V_{bb}=12\text{V}, I_L=7.5\text{A}, T_j=25^\circ\text{C}$ $V_{IN}=0, V_{bb}=12\text{V}, I_L=7.5\text{A}, T_j=150^\circ\text{C}$	- - - -	4.5 12 4.5 12	10 16 10 16	$\text{m}\Omega$
Output voltage drop limitation at small load currents(tab to pin 1,5)	$V_{ON(NL)}$	$T_j=-40 \sim 150^\circ\text{C}$	-	35	65	mV
Nominal load current(tab to pin1,5)	$I_{L(ISO)}$ $I_{L(NOM)}$	ISO Proposal ⁽⁶⁾ : $V_{ON}<=0.5\text{V}, T_C=85^\circ\text{C}, T_j<=150^\circ\text{C}$ SMD ⁽⁶⁾⁷⁾ : $V_{ON}<= 0.5\text{V}, T_a=85^\circ\text{C}, T_j<=150^\circ\text{C}$	33 10.5	41 12	- -	A
Turn-on time (to 90% V_{OUT})	t_{on}	$R_L=2.2\Omega, T_j = -40 \sim 150^\circ\text{C}$	-	160	400	us
Turn-off time (to 10% V_{OUT})	t_{off}	$R_L=2.2\Omega, T_j = -40 \sim 150^\circ\text{C}$	-	110	400	us
Slew rate on (25% to 50% V_{OUT})	dV / dt_{on}	$R_L=2.2\Omega, T_j = -40 \sim 150^\circ\text{C}$	-	0.2	0.5	V/us
Slew rate off (50% to 25% V_{OUT})	$-dV / dt_{off}$	$R_L=2.2\Omega, T_j = -40 \sim 150^\circ\text{C}$	-	0.3	0.65	V/us
Operating Parameters						
Operating Voltage($V_{IN}=0$)	$V_{bb(ON)}$	$T_j = -40 \sim 150^\circ\text{C}$	5.5		38	V
Under voltage shutdown ⁽⁷⁾⁸⁾	$V_{blN(u)}$		-	1.5	3.5	V
Under voltage restart of charge pump	$V_{bb(ucp)}$			3.7	5.5	V
Over voltage protection ⁽⁹⁾	$V_{Z,IN}$	$I_{bb}=15\text{mA}, T_j = -40 \sim 150^\circ\text{C}$	42.5	47.3	-	V
Standby current	$I_{bb(off)}$	$I_{IN}=0, T_j = -40 \sim 120^\circ\text{C}$ $I_{IN}=0, T_j = 150^\circ\text{C}$	- -	0.8 8	5.3 20	μA
Reverse Battery ⁽¹⁰⁾						
Reverse battery voltage	$-V_{bb}$		-	-	18	V
On-state resistance(Tab to pin 1,5) ⁽⁸⁾	$R_{ON(REV)}$	$V_{bb}=-8\text{V}, V_{IN}=0, I_L=-7.5\text{A}, R_{IS}=1\text{K}\Omega, T_j=25^\circ\text{C}$ $V_{bb}=-8\text{V}, V_{IN}=0, I_L=-7.5\text{A}, R_{IS}=1\text{K}\Omega, T_j=150^\circ\text{C}$ $V_{bb}=-12\text{V}, V_{IN}=0, I_L=-7.5\text{A}, R_{IS}=1\text{K}\Omega, T_j=25^\circ\text{C}$ $V_{bb}=-12\text{V}, V_{IN}=0, I_L=-7.5\text{A}, R_{IS}=1\text{K}\Omega, T_j=150^\circ\text{C}$	- - - -	9.5 15 5.5 12	13 22 10 16	$\text{m}\Omega$
Integrated resistor in V_{bb} line	R_{bb}	$I_s=1\text{mA}, V_{IN}=5\text{V} @ 125^\circ\text{C}$	65	95	125	Ω
Inverse operation ⁽¹¹⁾						
Output voltage drop(pin4 to pin 1, 5) ⁽⁸⁾	$-V_{ON(inv)}$	$I_L=-10\text{A}, R_{IS}=1\text{K}\Omega, T_j=25^\circ\text{C}$ $I_L=-10\text{A}, R_{IS}=1\text{K}\Omega, T_j=150^\circ\text{C}$	- -	800 600	- -	mV
Turn-on delay after inverse operation;	$t_{d(inv)}$	$I_L> 0\text{A}$ ⁽⁸⁾ $V_{IN(inv)}=V_{IN(fwd)} = 0\text{V}$	-	1	-	ms

Note:

5) Thermal resistance R_{thCH} case to heatsink (about 0.5 ... 0.9 K/W with silicone paste) not included!

6) Device on 76.2mm * 114mm * 1.57mm glass epoxyPCB. still air condition.

7) Not subject to production test, Parameters are calculated from Ron and R_{thjc} or R_{thja} .

8) not subject to production test, specified by design

9) See also $V_{ON(CL)}$ in circuit diagram page 8.

10) For operation at voltages higher than |16V| please see required schematic on page 9.

11) Permanent Inverse operation results eventually in a current flow via the intrinsic diode of the power DMOS. In this case the device switches on with a time delay $t_{d(inv)}$ after the transition from inverse to forward mode.

Electrical Characteristics

At $T_j=25^\circ\text{C}$, $V_{bb}=12\text{V}$ unless otherwise specified

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
Protection Functions¹²⁾						
Short circuit current limit (pin4 and tab to pin 1,5 at $V_{ON}=6\text{V}$ ^{13) 14)}	IL6(SC)	$T_j=-40^\circ\text{C}$ $T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	- - 65	100 90 80	130	A
Short circuit current limit (pin4 and tab to pin 1,5 at $V_{ON}=12\text{V}$ ¹³⁾	IL12(SC)	$T_j=-40^\circ\text{C}$ $T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	- - 40	80 70 60	120	A
Short circuit current limit (pin4 and tab to pin 1,5 at $V_{ON}=18\text{V}$ ^{13) 14)}	IL18(SC)	$T_j=-40^\circ\text{C}$ $T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	- - 33	63 55 47	81	A
Short circuit current limit (pin4 and tab to pin 1,5 at $V_{ON}=24\text{V}$ ¹³⁾	IL24(SC)	$T_j=-40^\circ\text{C}$ $T_j=25^\circ\text{C}$ $T_j=150^\circ\text{C}$	- - 18	43 38 33	65	A
Short circuit shutdown detection voltage	$V_{ON(SC)}$		2.5	3.5	4.5	V
Short circuit shutdown delay after input current positive slop Min value valid only if "off-signal" time exceeds 30us	$t_{d(SC1)}$	$V_{ON} > V_{ON(SC)}$, $T_j=-40 \sim 150^\circ\text{C}$	250	450	650	us
Output clamp(inductive load switch off) at $V_{OUT}= V_{bb}-V_{ON(CL)}$ (overvoltage) ¹⁵⁾	$V_{ON(CL)}$	$I_L=40\text{mA}$	38.5	42	-	V
Thermal overload trip temperature	T_{jt}		150	175	-	$^\circ\text{C}$
Thermal hysteresis	ΔT_{jt}		-	10	-	K

notes:

12) Integrated protection functions are designed to prevent IC destruction under fault conditions described in the data sheet. Fault conditions are considered as "outside" normal operating range. Protection functions are not designed for continuous repetitive operation.

13) Short circuit current limit for max. duration of $t_{d(SC1)}$, prior to shutdown, see also figures 3b on page 15.

14) not subject to production test, specified by design

15) See also figure 2b on page 14.

Electrical Characteristics

At $T_j=25^\circ\text{C}$, $V_{bb}=12\text{V}$ unless otherwise specified.

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit
Diagnostic Characteristics						
Current sense ratio, static on-condition KILIS = IL: IIS < IIS,min ¹⁶⁾ VIS < VOUT-5V , VbIN > 4.5V	KILIS	IL=30A, $T_j=-40^\circ\text{C}$ IL=30A, $T_j=25^\circ\text{C}$ IL=30A, $T_j=125^\circ\text{C}$ IL=7.5A, $T_j=-40^\circ\text{C}$ IL=7.5A, $T_j=25^\circ\text{C}$ IL=7.5A, $T_j=125^\circ\text{C}$ IL=2.5A, $T_j=-40^\circ\text{C}$ IL=2.5A, $T_j=25^\circ\text{C}$ IL=2.5A, $T_j=125^\circ\text{C}$ IL=0.5A, $T_j=-40^\circ\text{C}$ IL=0.5A, $T_j=25^\circ\text{C}$ IL=0.5A, $T_j=125^\circ\text{C}$ IIN=0(e.g. during deenergizing of inductive loads)	- 8000 8000 8000 5500 5500 5500 5000 5000 5000 4000 4000 4000	11000 11000 10500 10500 11000 11500 11000 10500 11000 12000 9900 11000 12000	- 13500 13000 13000 16500 17000 16500 17000 17500 18500 18000 18000 19000	- -
Sense current under fault conditions ¹⁷⁾	IIS,fault	$V_{ON} > 1\text{V}$,typ $T_j=-40 \sim 150^\circ\text{C}$	4	6.2	9	mA
Sense saturation current	IIS,lim	$V_{ON} < 1\text{V}$,typ $T_j=-40 \sim 150^\circ\text{C}$	4	7.5	9	mA
Fault-sense signal delay after input current positive slop	tdelay(fault)	$V_{ON} > 1\text{V}$,typ $T_j=-40 \sim 150^\circ\text{C}$	250	450	650	us
Current sense leakage current	IIS(LL)	IIN =0	-	0	0.5	uA
Current sense offset current	IIS(LH)	$V_{IN} =0$, $IL \leq 0$	-	0	2	uA
Minimum load current for sense functionality	IL(MIN)	$V_{IN} =0$, $T_j=-40 \sim 150^\circ\text{C}$	150	-	-	mA
Current sense settling time to IISstatic after input current positive slop ¹⁸⁾	tson(IS)	$IL =0 \rightarrow 20\text{A}$, $T_j=-40 \sim 150^\circ\text{C}$	-	300	650	us
Current sense settling time during on condition ¹⁸⁾	tslc(IS)	$IL =0 \rightarrow 20\text{A}$, $T_j=-40 \sim 150^\circ\text{C}$	-	50	100	us
Overvoltage protection	VZ,IS	$I_{bb}=15\text{mA}$, $T_j=-40 \sim 150^\circ\text{C}$	42.5	47.3	-	V
Input						
Required current capability of input switch	IIN(on)	$T_j=-40 \sim 150^\circ\text{C}$	-	1.5	3	mA
Input current for turn-off	IIN(off)	$T_j=-40 \sim 150^\circ\text{C}$	-	-	15	uA

notes:

16) See also figures 4.x and 6.x on page 15 and 16.

17) Fault conditions are overload during on (i.e. $V_{ON}>1\text{V}$ typ.), overtemperature and short circuit; see also truth table on page 7.

18) not subject to production test, specified by design

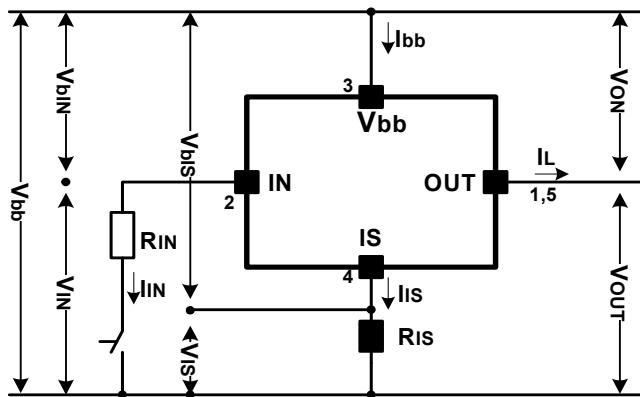
Application Information

1. Truth Table

Sense current under fault conditions	Input Current Level	Output Level	Current Sense
Normal operation	L H	L H	= 0 (IIS(LL)) nominal
Overload ¹⁹⁾	L H	L H	= 0 (IIS(LL)) IIS,fault
Short circuit to GND ²⁰⁾	L H	L L	= 0 (IIS(LL)) IIS,fault
Overttemperature	L H	L L	= 0 (IIS(LL)) IIS,fault
Short circuit to V _{bb}	L H	H H	= 0 (IIS(LL)) < nominal ²¹⁾
Open load	L H	Z H	= 0 (IIS(LL)) = 0 (IIS(LH))

L = "LOW" Level, Z = High impedance, potential depends on external circuit, H = "HIGH" Level

2.Terms



Notes:

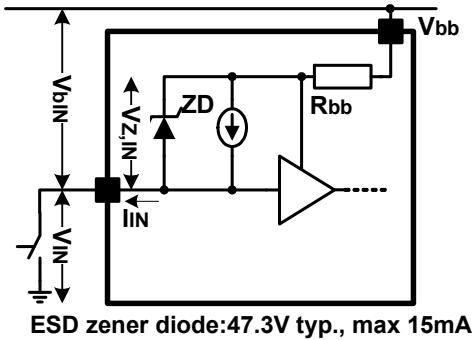
19) Overload is detected at the following condition: 1V (typ.) < VON < 3.5V (typ.). See also page 10.

20) Short Circuit is detected at the following condition: VON > 3.5V (typ.). See also page 11.

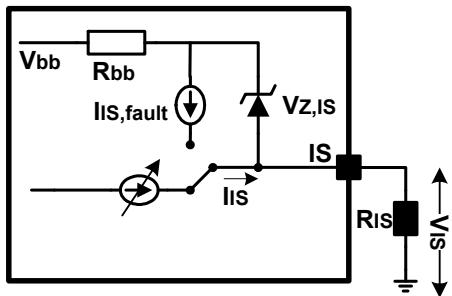
21) Low ohmic short to V_{bb} may reduce the output current IL and therefore also the sense current IIS.

3. Detailed Function Blocks

3.1 Input circuit(ESD protection)



3.2 Current sense output

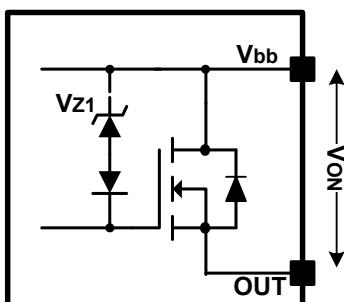


$V_{Z,IS} = 47.3$ V (typ.), $R_{IS} = 1\text{K}\Omega$ nominal (or $1\text{K}\Omega /n$, if n devices are connected in parallel). $I_S = I_L/k_{ilis}$ can be only driven by the internal circuit as long as $V_{OUT} - V_{IS} > 5$ V. Therefore R_{IS} should be less than

$$\frac{V_{bb} - 5}{9 \text{ mA}}$$

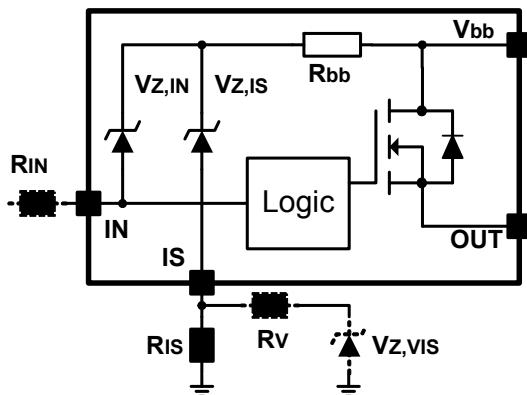
Note: For large values of R_{IS} the voltage V_{IS} can reach almost V_{bb} . See also overvoltage protection. If you don't use the current sense output in your application, you can leave it open.

3.3 Inductive and overvoltage output clamp



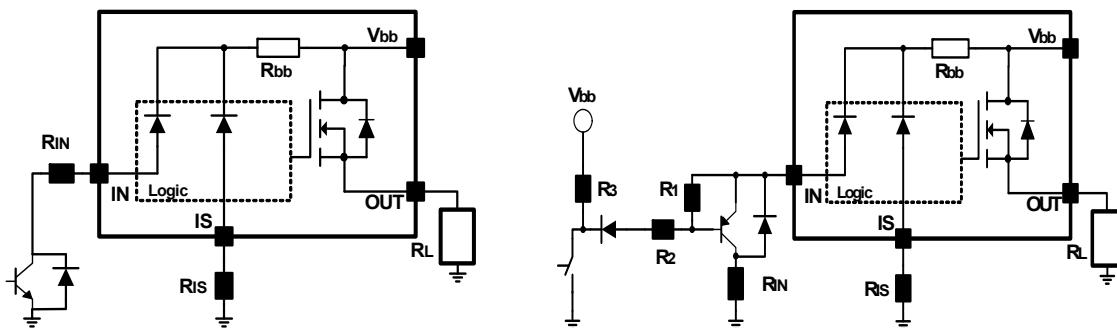
V_{ON} is clamped to $V_{ON(CL)} = 42$ V typ

3.4 Overvoltage protection of logic part



$R_{bb} = 95\Omega$ typ., $V_{Z,IN} = V_{Z,IS} = 47.3V$ typ., $R_{IS} = 1K\Omega$ nominal. Note that when overvoltage exceeds 47.3V typ. a voltage above 5V can occur between IS and GND, if R_V , $V_{Z,VIS}$ are not used.

3.5 Reverse battery protection

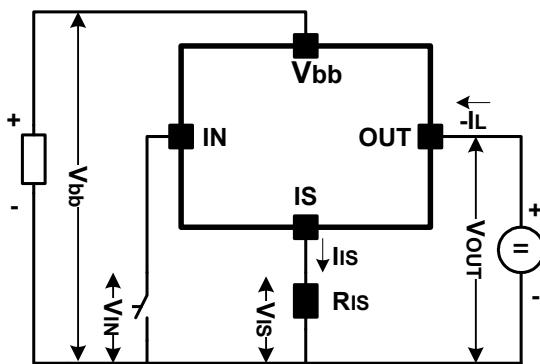


R_{IS} typ. $1K\Omega$. Add R_{IN} for reverse battery protection in application with V_{bb} above 16V;

$$\frac{1}{R_{IN}} + \frac{1}{R_{IS}} = \frac{0.082A}{|V|_{bb} - 9V}$$

To minimise power dissipation at reverse battery operation, the overall current into the IN and IS pin should be about 82mA. The current can be provided by using a small signal diode D in parallel to the input switch, by using a MOSFET input switch or by proper adjusting the current through R_{IS} . Since the current via R_{bb} generates additional heat in the device, this has to be taken into account in the overall thermal consideration.

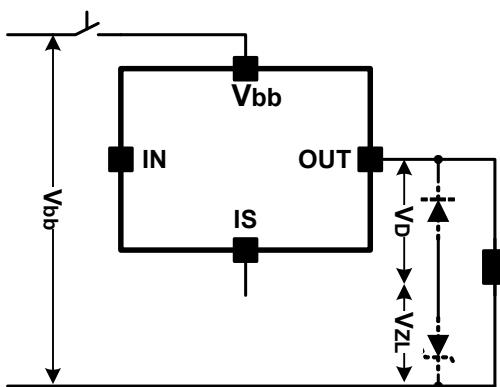
3.6 Inverse load current operation



The device can be operated in inverse load current mode ($V_{OUT} > V_{bb} > 0V$). The current sense feature is not available during this kind of operation ($I_{IS} = 0$). In case of inverse operation the intrinsic drain source diode is eventually conducting resulting in considerably increased power dissipation. The transition from inverse to forward mode can result in a delayed switch on.

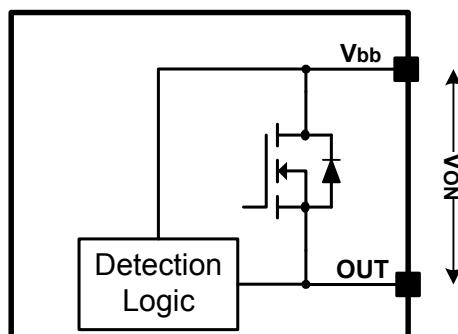
Note: Temperature protection during inverse load current operation is not possible!

3.7 V_{bb} disconnect with energised inductive load



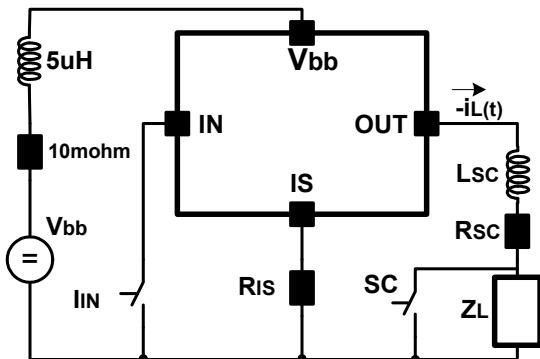
Provide a current path with load current capability by using a diode, a Z-diode, or a varistor. ($V_{ZL} + V_D < 38V$ if $R_{IN} = 0$). For higher clamp voltages currents at IN and IS have to be limited to 120 mA.

3.8 Overload detection



Fault Condition: $V_{ON} > 1V$ typ.

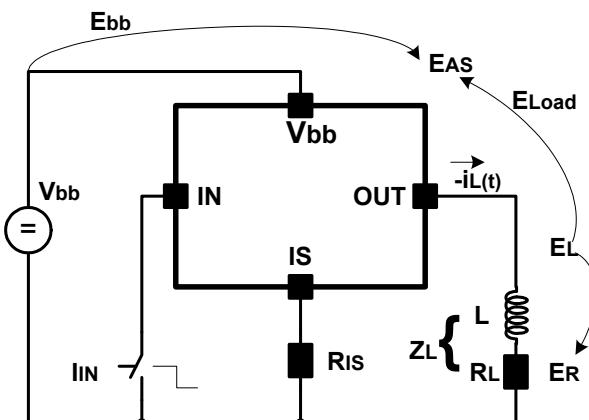
3.9 Short circuit



Fault Condition: $V_{ON} > V_{ON(SC)}$ (3.5V typ.) and $t > t_d(SC)$ (typ. 450us)

Short circuit is a combination of primary and secondary impedance's and a resistance's. Allowable combinations of minimum, secondary resistance for full protection at given secondary inductance and supply voltage for single short circuit event:

3.10 Inductive load switch-off energy dissipatoion



Energy stored in load inductance:

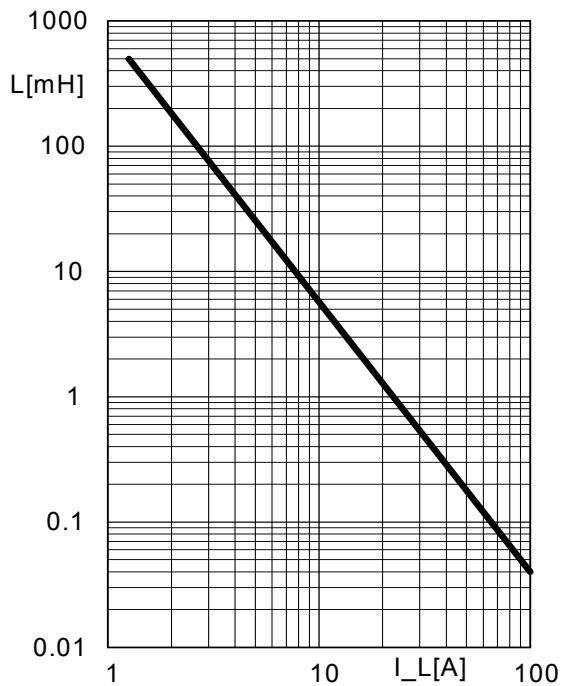
$$E_L = 1/2 \cdot L \cdot I_L^2$$

While demagnetizing load inductance, the energy dissipated in MOSFET is

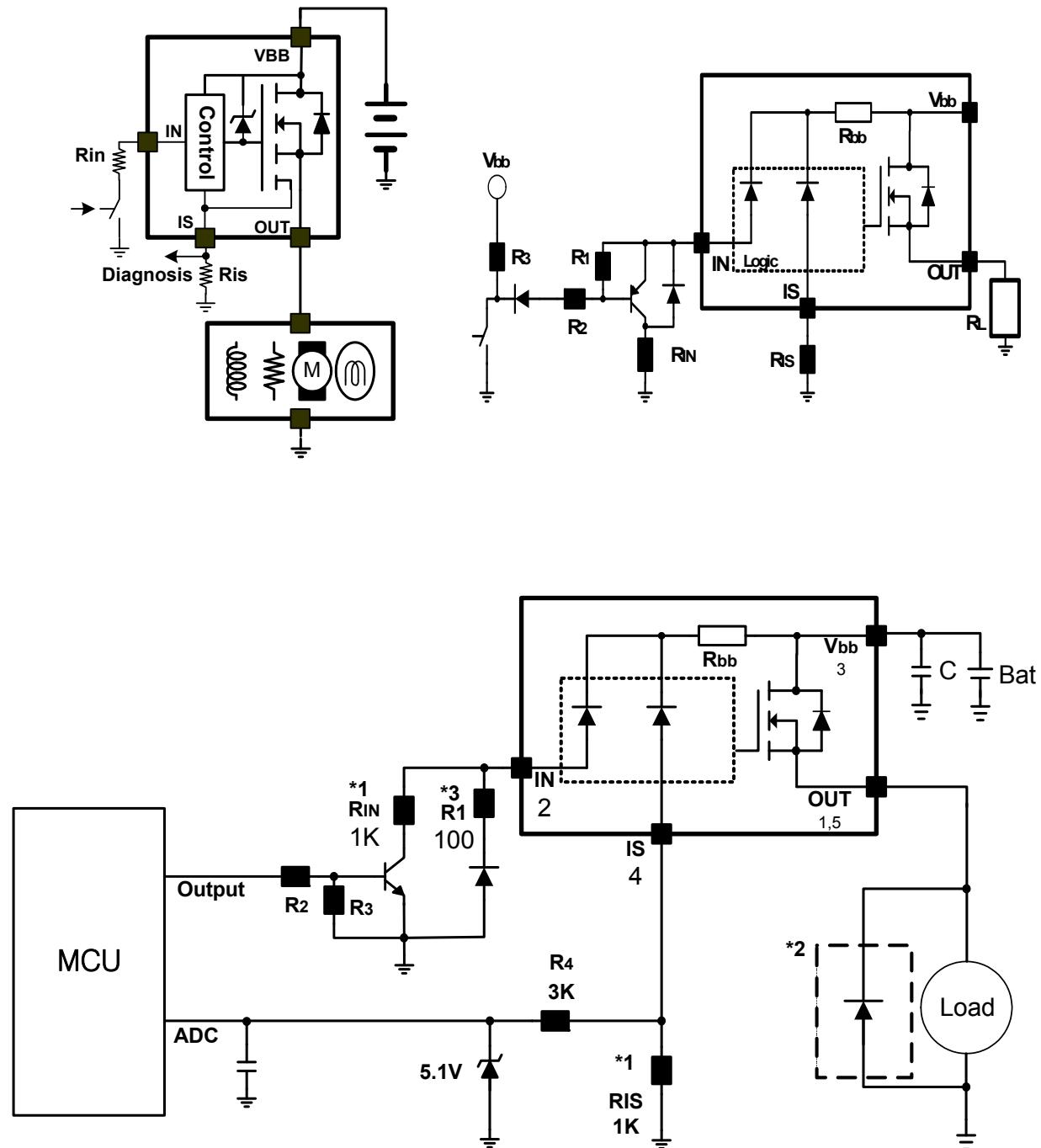
$$E_{AS} = E_{bb} + E_L - E_R = \int V_{ON(CL)} \cdot i_L(t) dt$$

with an approximate solution for $R_L > 0$, $V_{out(CL)} = V_{ON(CL)} - V_{bb}$:

$$E_{AS} = \frac{I_L \cdot L}{2 \cdot R_L} (V_{ON(CL)}) \ln \left(1 + \frac{I_L \cdot R_L}{|V_{OUT(CL)}|} \right)$$

3.11 Maximum allowable load inductance for a single switch off $L=f(I_L)$; $T_j = 150^\circ\text{C}$, $V_{bb}=12\text{V}$, $R_L = 0\Omega$ 

Typical Application Circuit



1) R_{IS} and R_{IN} is recommended as 1k

2) Put diode or capacitor between load to protect device or to remove noise.

3) For reverse battery protection function, R₁ should be used less than 120Ω at -18V.

Timing Diagrams

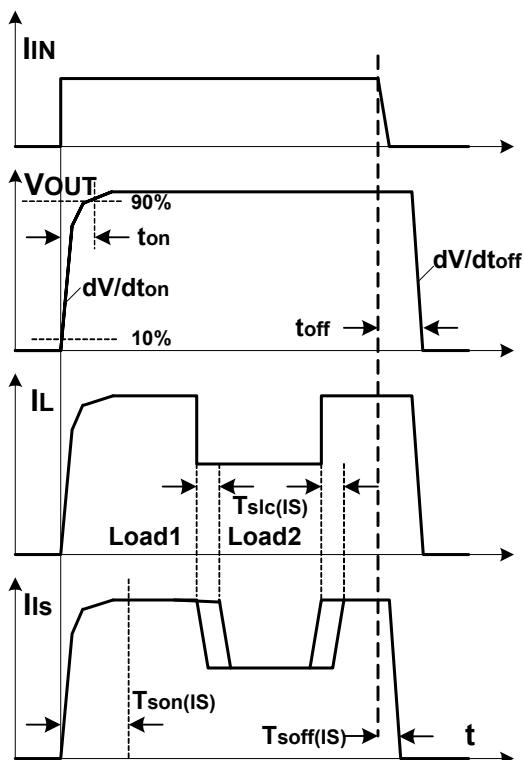


Figure 1a. Switching a resistive load, change of load current in on-condition

The sense signal is not valid during a settling time after turn-on/off and after change of load current

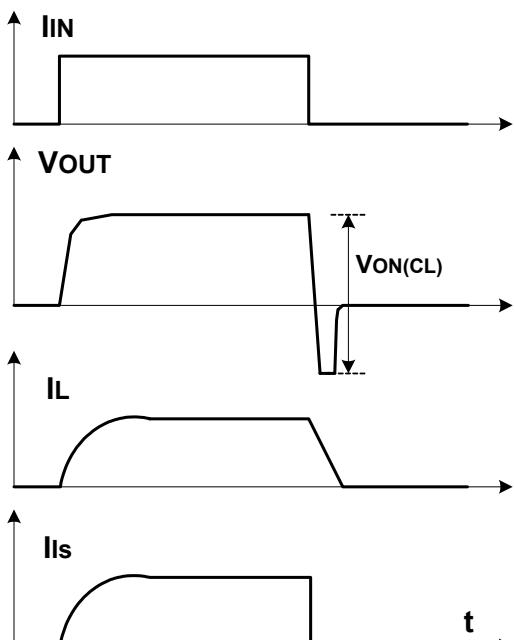


Figure 2b. Switching an inductive load

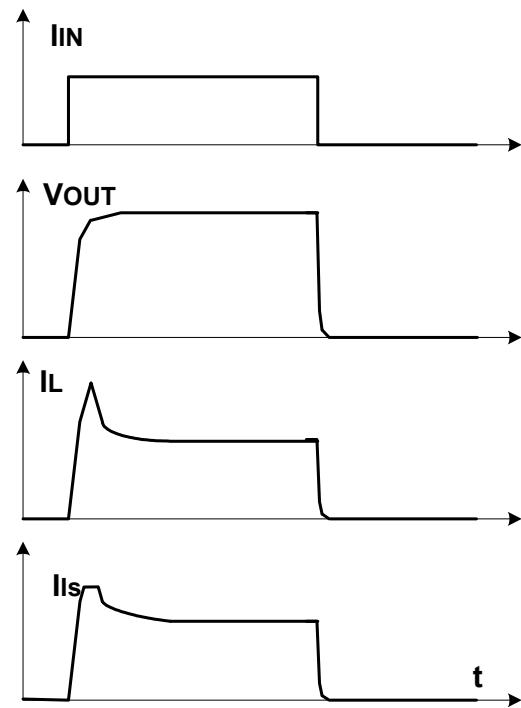


Figure 2a. Switching motors and lamps

As long as $V_{BLS} < V_{Z,IS}$ the sense current will never exceed $I_{IS,fault}$ and/or $I_{IS,lim}$

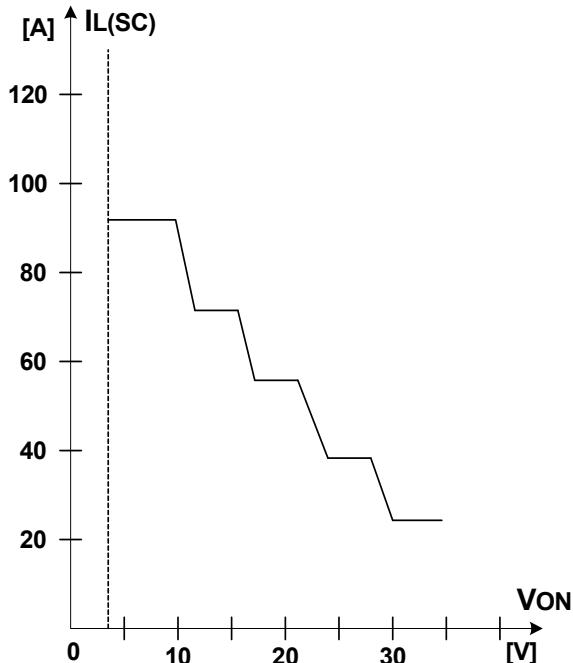


Figure 3a. Typ. current limitation characteristic

In case of $V_{ON} > V_{ON(SC)}$ (typ. 3.5V) the device will be switched off by internal short circuit detection

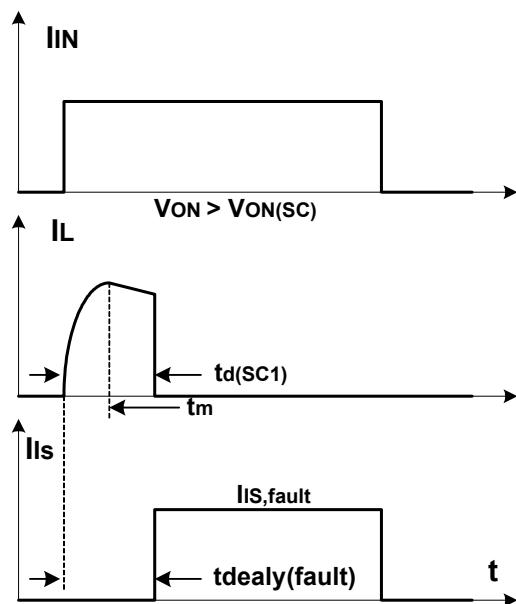


Figure 3b. Short circuit type one:
shut down by short circuit detection, reset by $I_{IN}=0$

Shut down remains latched until next reset via input

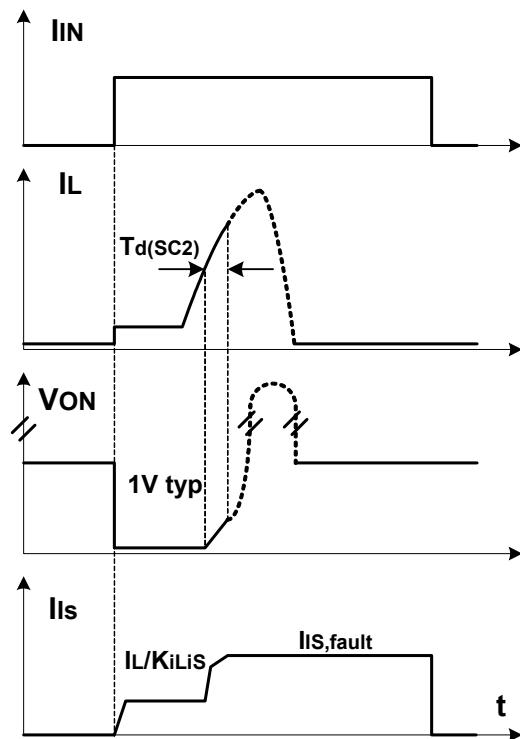


Figure 3c. Short circuit type two:
shut down by short circuit detection, reset by $I_{IN}=0$

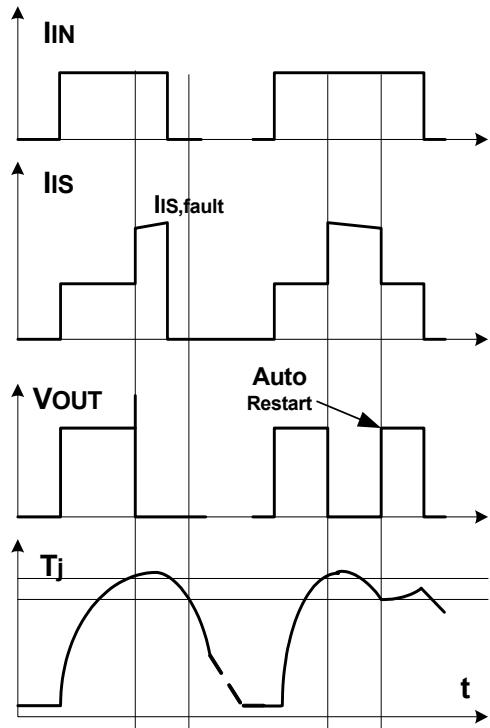


Figure 4a. Overtemperature Reset if $T_j < T_{jt}$

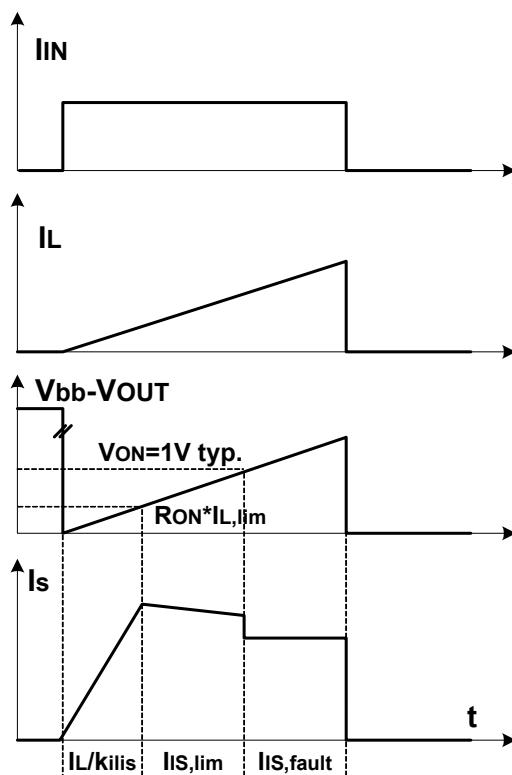


Figure 4b. Overload Reset if $T_j < T_{jt}$

3.VB Drop Voltage Diagram

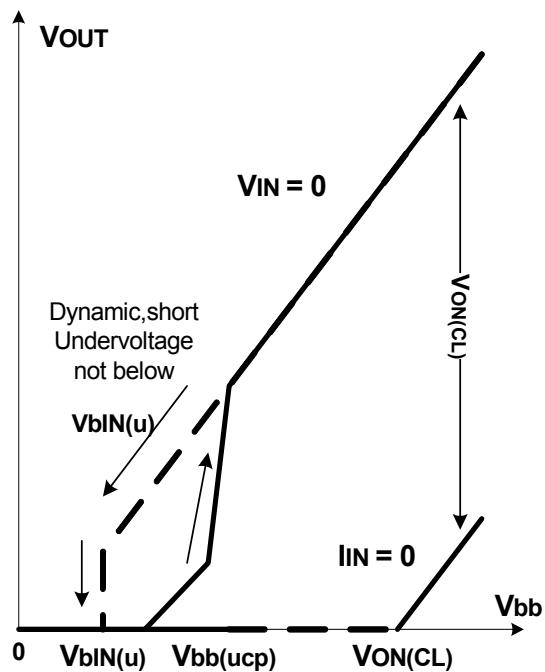


Figure 5. Undervoltage restart of charge pump overvoltage clamp

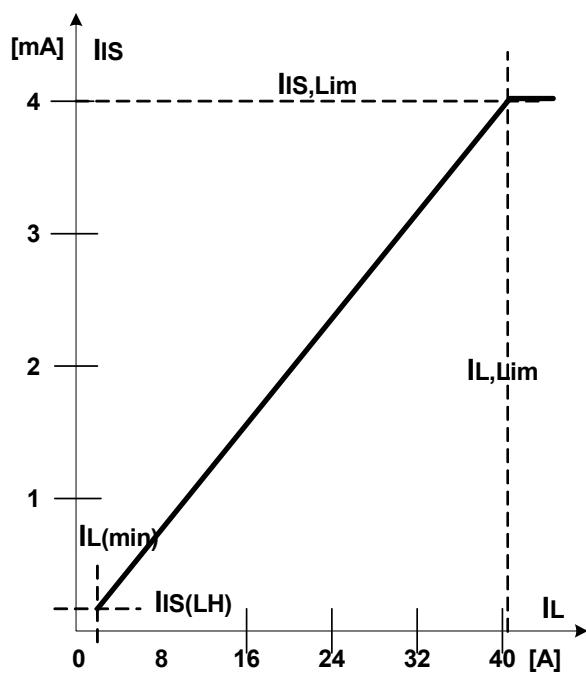


Figure 6a. Current sense versus load current

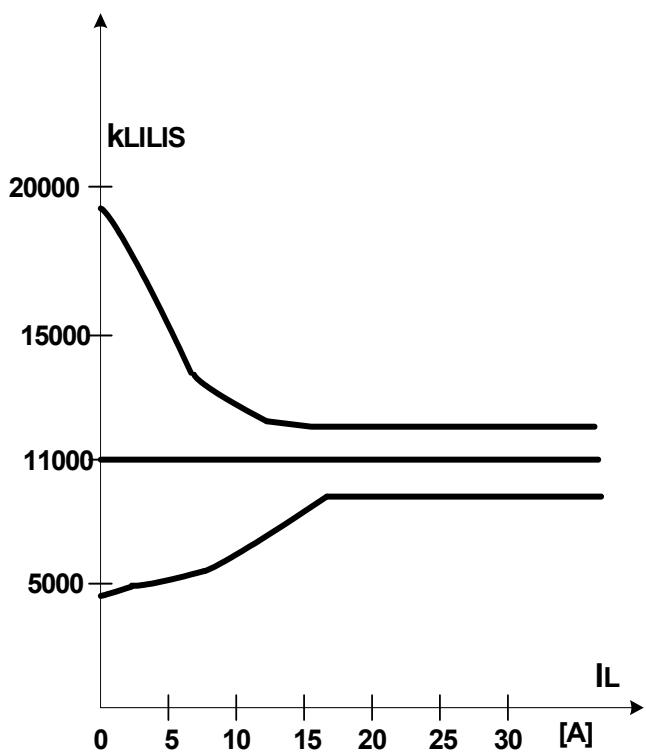


Figure 6b. Current sense ratio

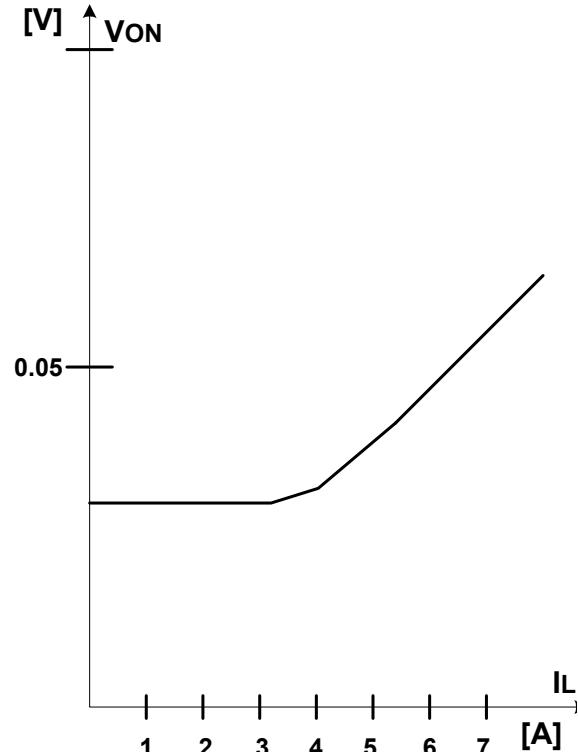
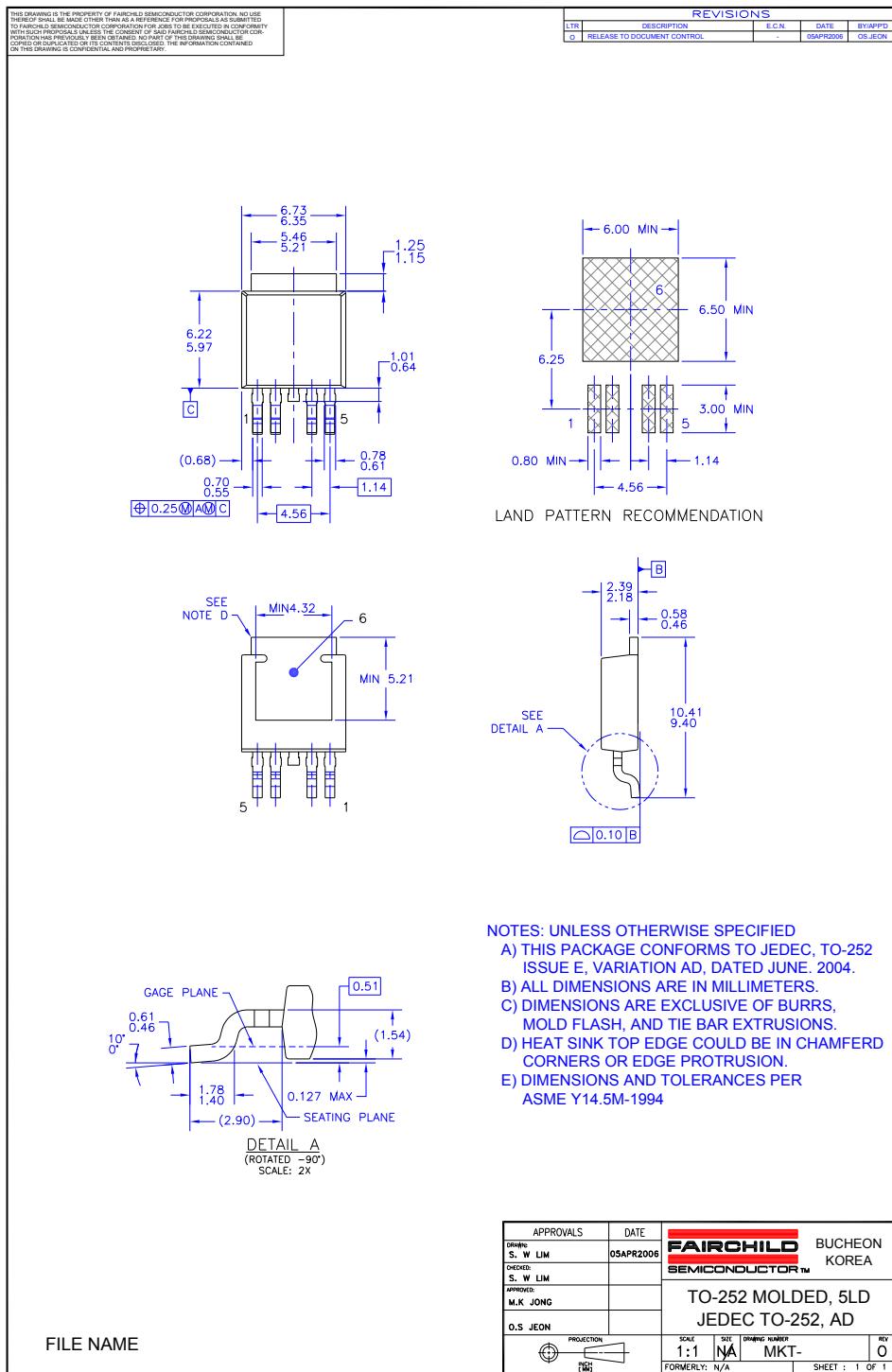


Figure 7. Output voltage drop versus load current

Package Dimensions



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E ² CMOS™	HiSeC™	MSXPro™	RapidConfigure™	TruTranslation™
EnSigna™	I ² C™	OCX™	RapidConnect™	UHC™
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The Power Franchise™		OPTOPLANAR™	SPM™	
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